3.3V, High-Speed, RS-485/RS-422 Transceiver in SOT Package

General Description

The MAX3362 low-power, high-speed transceiver for RS-485/RS-422 communication operates from a single +3.3V power supply. The device contains one differential transceiver consisting of a line driver and receiver. The transceiver operates at data rates up to 20Mbps, with an output skew of less than 6ns. Driver and receiver propagation delays are guaranteed below 50ns. This fast switching and low skew make the MAX3362 ideal for multidrop clock/data distribution applications.

The output level is guaranteed at +1.5V with a standard 54Ω load, compliant with RS-485 specifications. The transceiver draws 1.7mA supply current when unloaded or fully loaded with the drivers disabled. Additionally, the MAX3362 has a low-power shutdown mode, reducing the supply current to 1µA.

The MAX3362 has a 1/8-unit-load receiver input impedance, allowing up to 256 transceivers on the bus. The MAX3362 is designed for half-duplex communication. The device has a hot-swap feature that eliminates false transitions on the data cable during circuit initialization. The drivers are short-circuit current limited, and a thermal shutdown circuit protects against excessive power dissipation.

The MAX3362 is available in an 8-pin SOT package and specified over industrial and automotive temperature ranges.

Applications

Clock/Data Distribution
Telecom Equipment
Security Equipment
Point-of-Sale Equipment
Industrial Controls

Pin Configuration and Functional Diagram appear at end of data sheet.

Benefits and Features

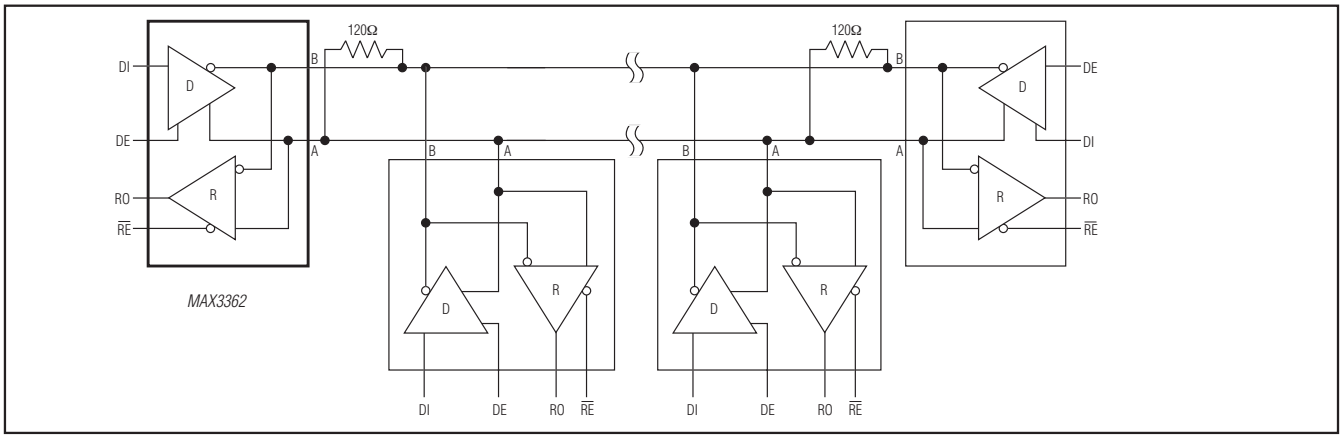
- Ideal for Multidrop RS-485/RS-422 Clock/Data Distribution
  - Guaranteed 20Mbps Data Rate
  - 6ns (max) Transmitter and Receiver Skew
  - -7V to +12V Common-Mode Range
  - Hot-Swap Feature
  - Allows Up to 256 Transceivers on the Bus
  - Half-Duplex Operation
  - Interoperable with +5V Logic
- Low Power Consumption Minimizes Thermal Dissipation
  - Operates from a Single +3.3V Supply
  - 1.7mA Operating Supply Current
  - 1µA Low-Power Shutdown Mode
- Space-Saving 8-Pin SOT Package
- Protection Features and Wide Temperature Range Increase System Reliability in Harsh Environments
  - Current Limiting and Thermal Shutdown
  - Automotive Temperature Range Variants

Ordering Information

PART	TEMP RANGE	PIN-PACKAGE	TOP MARK	PKG CODE
MAX3362EKA-T	-40°C to +85°C	8 SOT23-8	AAJL	S8-1
MAX3362AKA-T	-40°C to +125°C	8 SOT23-8	AALL	S8-1
MAX3362EKA#T	-40°C to +85°C	8 SOT23-8	#AEPH	S8-1
MAX3362AKA#T	-40°C to +125°C	8 SOT23-8	#AEPP	S8-1

#Indicates an RoHS-compliant part.

Typical Operating Circuit



maxim integrated™

Absolute Maximum Ratings

All voltages with respect to GND.
V_{CC}, $\overline{RE}$, DE, DI .....-0.3V to +6V
Receiver Input Voltages, Driver Output Voltages (A, B) .....-8V to +13V
Receiver Input Current, Driver Output Current (A, B).....250mA
$|V_A - V_B|$.....+8V
Receiver Output Voltage (RO).....-0.3V to (V_{CC} + 0.3V)

Continuous Power Dissipation (T_A = +70°C)
8-Pin SOT (derate 9.7mW/°C above +70°C) 777mW
Operating Temperature Range
MAX3362E__ ..... -40°C to +85°C
MAX3362A__ ..... -40°C to +125°C
Storage Temperature Range .....-65°C to +150°C
Junction Temperature.....+150°C
Lead Temperature (soldering, 10s) .....+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

DC Electrical Characteristics

(V_{CC} = +3.3V ±5%, T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at V_{CC} = +3.3V and T_A = +25°C.) (Notes 1, 2)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
DRIVER						
Differential Driver Output	V_{OD}	Figure 1, R_L = 100Ω (RS-422) (extended temperature range)	2.0			V
		Figure 1, R_L = 100Ω (automotive temperature range)	1.5			
		Figure 1, R_L = 54Ω (RS-485) (extended temperature range)	1.5			
Change in Magnitude of Differential Output Voltage	ΔV_{OD}	Figure 1, R_L = 54Ω or 100Ω (Note 3)			0.2	V
Driver Common-Mode Output Voltage	V_{OC}	Figure 1, R_L = 54Ω or 100Ω			3	V
Change In Magnitude of Common-Mode Voltage	ΔV_{OC}	Figure 1, R_L = 54Ω or 100Ω (Note 3)			0.2	V
Input High Voltage	V_{IH}	DE, DI, $\overline{RE}$	2.0			V
Input Low Voltage	V_{IL}	DE, DI, $\overline{RE}$			0.8	V
Input Hysteresis	V_{HYS}	DE, DI, $\overline{RE}$		50		mV
Input Current (DE, DI, $\overline{RE}$)	I_{IN}	0 ≤ V_{IN} ≤ 5V			±1	µA
Driver Short-Circuit Output Current	I_{OSD}	0 ≤ V_{OUT} ≤ 12V (Note 4)			+250	mA
		-7V ≤ V_{OUT} ≤ V_{CC} (Note 4)	-250			
Driver Short-Circuit Foldback Output Current	I_{OSDF}	(V_{CC} - 1V) ≤ V_{OUT} ≤ 12V (Note 4)	+25			mA
		-7V ≤ V_{OUT} ≤ 1V (Note 4)			-25	

DC Electrical Characteristics (continued)

(V_{CC} = +3.3V ±5%, T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at V_{CC} = +3.3V and T_A = +25°C.) (Notes 1, 2)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Thermal Shutdown Threshold	V_{TS}				150		°C
Thermal Shutdown Hysteresis	V_{TSH}				10		°C
RECEIVER							
Receiver Differential Threshold Voltage	V_{TH}	-7V ≤ V_{CM} ≤ 12V		-200	0	+200	mV
Receiver Input Hysteresis	ΔV_{TH}	V_A + V_B = 0			25		mV
Receiver Output High Voltage	V_{OH}	I_O = -1mA, V_A - V_B = V_{TH}		V_{CC} - 0.4			V
Receiver Output Low Voltage	V_{OL}	I_O = 1mA, V_A - V_B = -V_{TH}				0.4	V
Three-State Output Current at Receiver	I_{OZR}	0 ≤ V_O ≤ V_{CC}				±1	µA
Receiver Input Resistance	R_{IN}	V_{CM} = 12V		96			kΩ
Receiver Input Current	I_{IN}	DE = GND, V_{CC} = GND or 3.465V	V_{IN} = +12V			125	µA
			V_{IN} = -7V	-100			
Receiver Output Short-Circuit Current	I_{OSR}	0 ≤ V_{RO} ≤ V_{CC}				±150	mA
POWER SUPPLY							
Supply Voltage	V_{CC}			3.135	3.300	3.465	V
Supply Current in Normal Operation (Static Condition)	I_Q	No load, DI = V_{CC} or GND			1.7	3	mA
Supply Current in Shutdown Mode	I_{SHDN}	DE = GND, $\overline{RE}$ = V_{CC}			1	10	µA

Switching Characteristics

(V_{CC} = +3.3V ±5%, T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at V_{CC} = +3.3V and T_A = +25°C.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Driver Propagation Delay	t_{PDLH}	Figures 2 and 3, R_L = 54Ω, C_L = 50pF			50	ns
	t_{PDHL}				50	
Driver Differential Output Rise or Fall Time	t_{DR}	Figures 2 and 3, R_L = 54Ω, C_L = 50pF			12.5	ns
	t_{DF}				12.5	
Driver Output Skew	t_{DSKEW}	Figures 2 and 3, R_L = 54Ω, C_L = 50pF, t_{DSKEW} = $\lvert t_{PDLH} - t_{PDHL} \rvert$			6	ns
Maximum Data Rate	f_{MAX}		20			Mbps
Driver Enable to Output Low	t_{PDZL}	Figure 4, R_L = 500Ω, C_L = 50pF			100	ns
Driver Disable Time from Low	t_{PDLZ}	Figure 4, R_L = 500Ω, C_L = 50pF			100	ns
Driver Disable Time from High	t_{PDHZ}	Figure 5, R_L = 500Ω, C_L = 50pF			100	ns

Switching Characteristics (continued)

(V_{CC} = +3.3V ±5%, T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at V_{CC} = +3.3V and T_A = +25°C.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Driver Enable to Output High	t_{PDZH}	Figure 5, R_L = 500Ω, C_L = 50pF			100	ns
Receiver Propagation Delay	t_{PRLH}	Figure 6, C_L = 15pF			50	ns
	t_{PRHL}				50	
Receiver Output Skew	t_{RSKEW}	Figure 6, C_L = 15pF $t_{RSKEW} = \lvert t_{PRLH} - t_{PRHL} \rvert$			6	ns
Receiver Enable to Output Low	t_{PRZL}	Figure 7, R_L = 1kΩ, C_L = 15pF			100	ns
Receiver Enable to Output High	t_{PRZH}	Figure 7, R_L = 1kΩ, C_L = 15pF			100	ns
Receiver Disable Time from Low	t_{PRLZ}	Figure 7, R_L = 1kΩ, C_L = 15pF			100	ns
Receiver Disable Time from High	t_{PRHZ}	Figure 7, R_L = 1kΩ, C_L = 15pF			100	ns
Time to Shutdown	t_{SD}	(Note 5)	50		600	ns
Driver Enable from Output High to Shutdown	t_{PDHS}		50		600	ns
Driver Enable from Output Low to Shutdown	t_{PDLS}		50		600	ns
Receiver Enable from Output High to Shutdown	t_{PRHS}		50		600	ns
Receiver Enable from Output Low to Shutdown	t_{PRLS}		50		600	ns
Time to Normal Operation	t_{NO}	(Note 6)		1500	3000	ns
Driver Enable from Shutdown to Output High	t_{PDSH}	Figure 5 R_L = 500Ω, C_L = 50pF		1500	3000	ns
Driver Enable from Shutdown to Output Low	t_{PDSL}	Figure 4 R_L = 500Ω, C_L = 50pF		1500	3000	ns
Receiver Enable from Shutdown to Output High	t_{PRSH}	Figure 7 R_L = 1kΩ, C_L = 15pF		1500	3000	ns
Receiver Enable from Shutdown to Output Low	t_{PRSL}	Figure 7 R_L = 1kΩ, C_L = 15pF		1500	3000	ns

Note 1: Devices production tested at +25°C. Over-temperature limits are guaranteed by design.

Note 2: All currents into the device are positive; all currents out of the device are negative. All voltages are referenced to device ground, unless otherwise noted.

Note 3: ΔV_{OD} and ΔV_{OC} are the changes in V_{OD} and V_{OC}, respectively, when the DI input changes state.

Note 4: The short-circuit output current applies to peak current just prior to foldback-current limiting; the short-circuit foldback output current applies during current limiting to allow a recovery from bus contention.

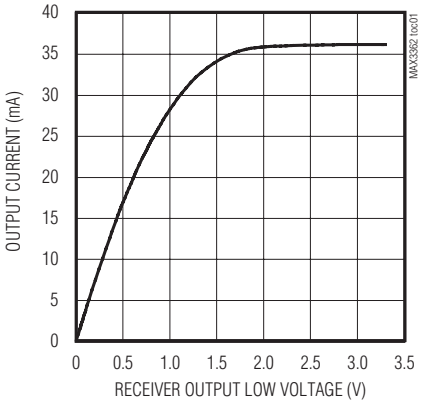
Note 5: Shutdown is enabled by bringing $\overline{RE}$ high and DE low. If the enable inputs are in this state for less than 50ns, the device is guaranteed not to enter shutdown. If the enable inputs are in this state for at least 600ns, the device is guaranteed to have entered shutdown.

Note 6: Transition time from shutdown mode to normal operation.

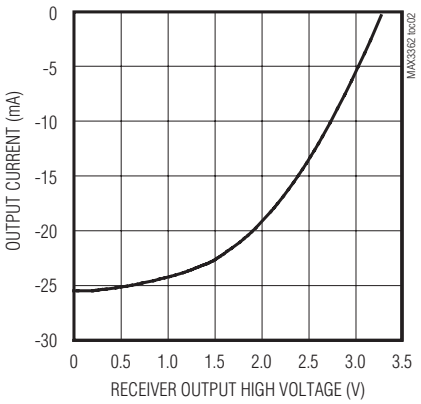
Typical Operating Characteristics

(V_{CC} = +3.3V, T_A = +25°C, unless otherwise noted.)

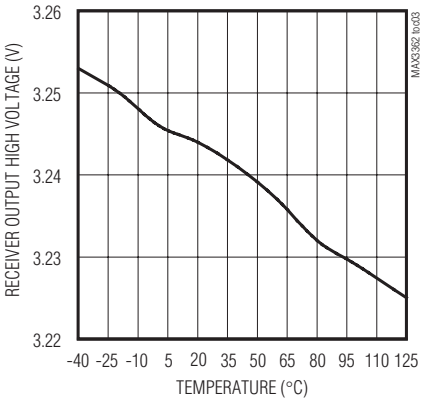
OUTPUT CURRENT vs. RECEIVER OUTPUT LOW VOLTAGE



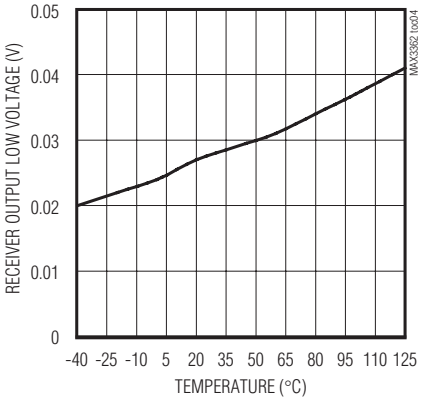
OUTPUT CURRENT vs. RECEIVER OUTPUT HIGH VOLTAGE



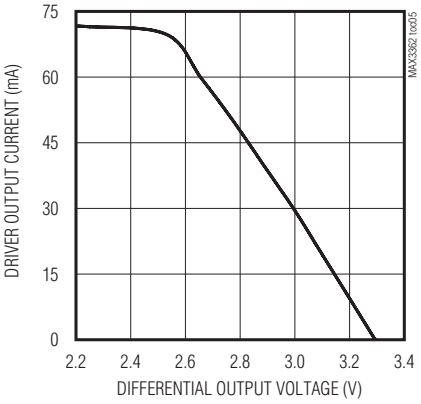
RECEIVER OUTPUT HIGH VOLTAGE vs. TEMPERATURE



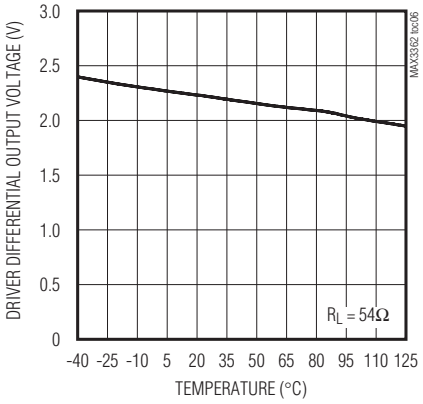
RECEIVER OUTPUT LOW VOLTAGE vs. TEMPERATURE



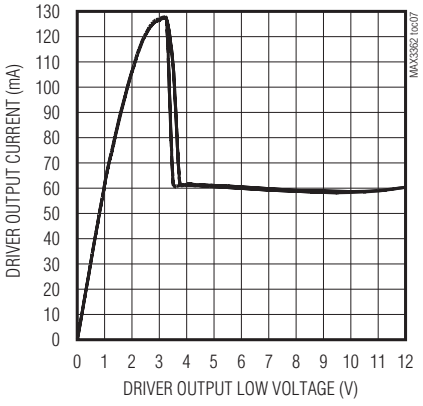
DRIVER OUTPUT CURRENT vs. DIFFERENTIAL OUTPUT VOLTAGE



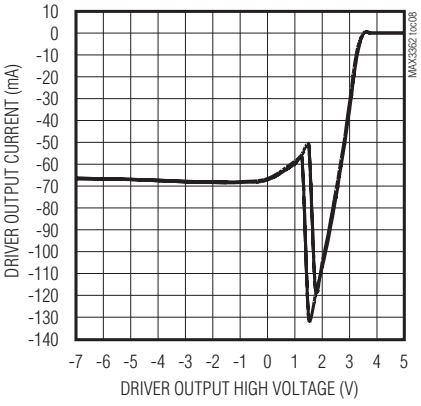
DRIVER DIFFERENTIAL OUTPUT VOLTAGE vs. TEMPERATURE



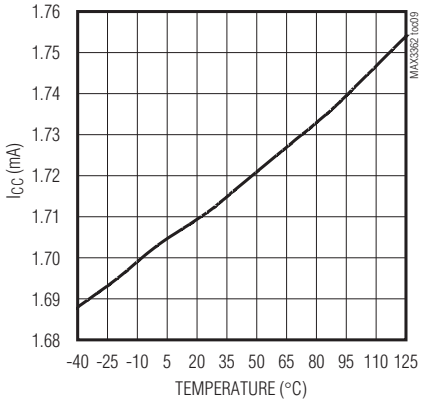
DRIVER OUTPUT CURRENT vs. DRIVER OUTPUT LOW VOLTAGE



DRIVER OUTPUT CURRENT vs. DRIVER OUTPUT HIGH VOLTAGE



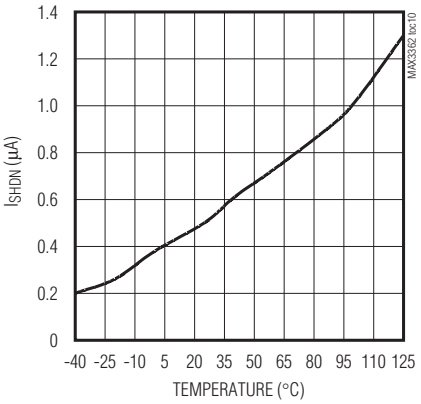
SUPPLY CURRENT vs. TEMPERATURE



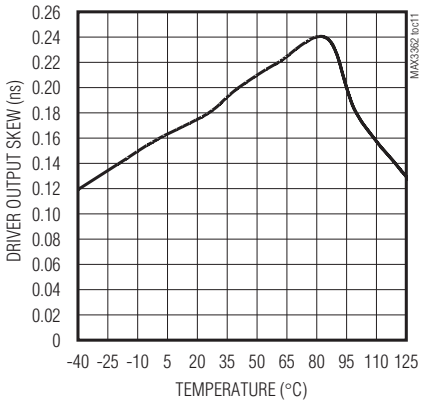
Typical Operating Characteristics (continued)

(V_{CC} = +3.3V, T_A = +25°C, unless otherwise noted.)

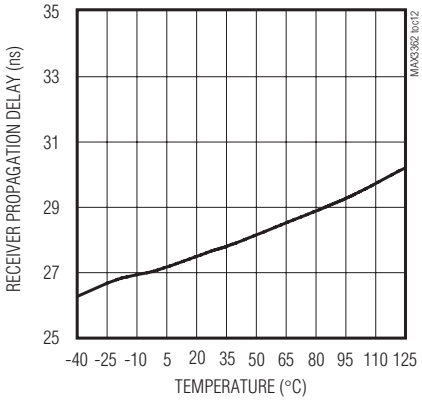
SHUTDOWN SUPPLY CURRENT vs. TEMPERATURE



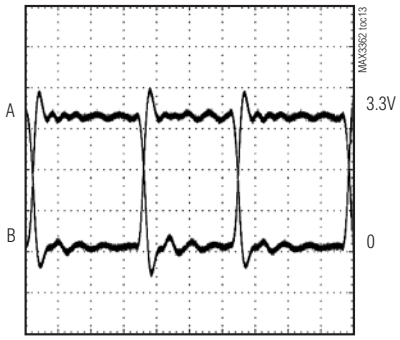
DRIVER OUTPUT SKEW vs. TEMPERATURE



RECEIVER PROPAGATION DELAY vs. TEMPERATURE

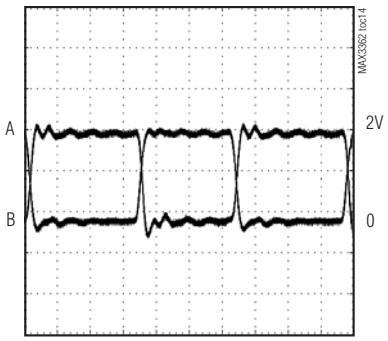


UNLOADED DRIVER OUTPUT WAVEFORM



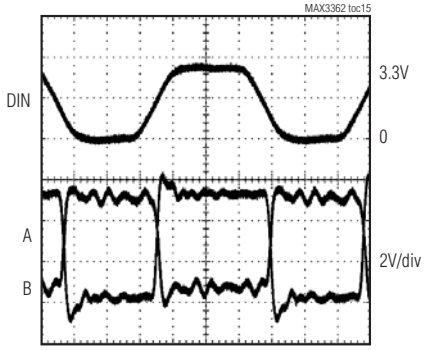
f_{DIN} = 16Mbps

LOADED DRIVER OUTPUT WAVEFORM



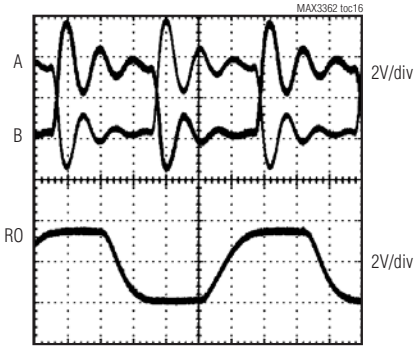
f_{DIN} = 16Mbps
R_L = 54Ω

DRIVER PROPAGATION DELAY



f_{DIN} = 16Mbps

RECEIVER PROPAGATION DELAY



f_{DIN} = 16Mbps

Pin Description

PIN	NAME	DESCRIPTION
1	RO	Receiver Output. RO is high if the receiver input differential (A-B) ≥ 200mV and the receiver is enabled ($\overline{RE}$ is low). RO is low if the receiver input differential (A-B) ≤ -200mV and the receiver is enabled.
2	$\overline{RE}$	Receiver Output Enable. Driving $\overline{RE}$ low enables RO. RO is high impedance when $\overline{RE}$ is high. Drive $\overline{RE}$ high and DE low (disable both receiver and driver outputs) to enter low-power shutdown mode.
3	DE	Driver Output Enable. Driving DE high enables driver outputs. These outputs are high impedance when DE is low. Drive $\overline{RE}$ high and DE low (disable both receiver and driver outputs) to enter low-power shutdown mode.
4	DI	Driver Input. Driving DI low forces the noninverting output low and inverting output high, when the driver is enabled (DE is high). Driving DI high forces the noninverting output high and inverting output low.
5	GND	Ground
6	A	Noninverting Receiver Input and Noninverting Driver Output
7	B	Inverting Receiver Input and Inverting Driver Output
8	V_{CC}	Supply Voltage. V_{CC} = 3.3V ±5%. Bypass V_{CC} to GND with a 0.1µF capacitor.

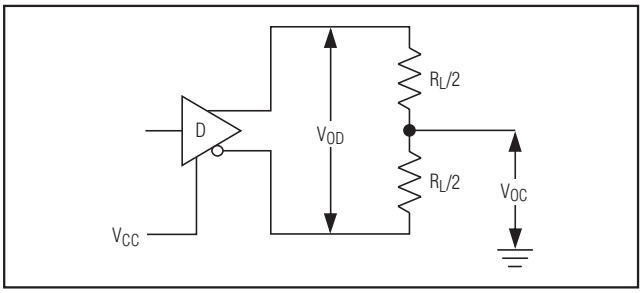


Figure 1. Driver DC Test Load

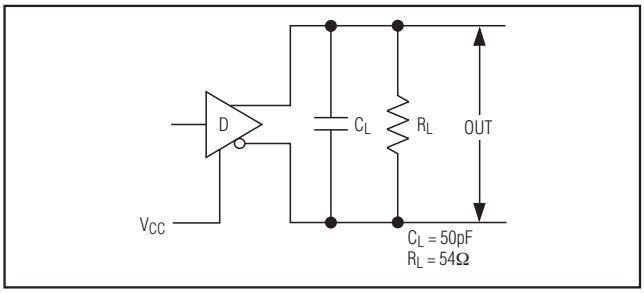


Figure 2. Driver Timing Test Circuit

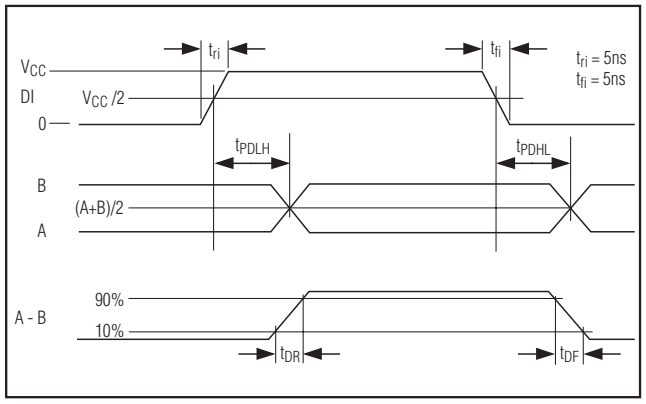


Figure 3. Driver Propagation Delay

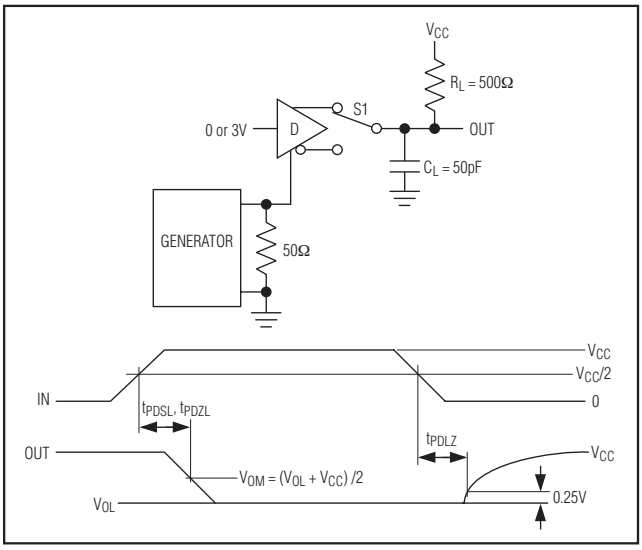


Figure 4. Driver Enable and Disable Times (t_{PDSL}, t_{PDZL}, t_{PDLS}, t_{PDLZ})

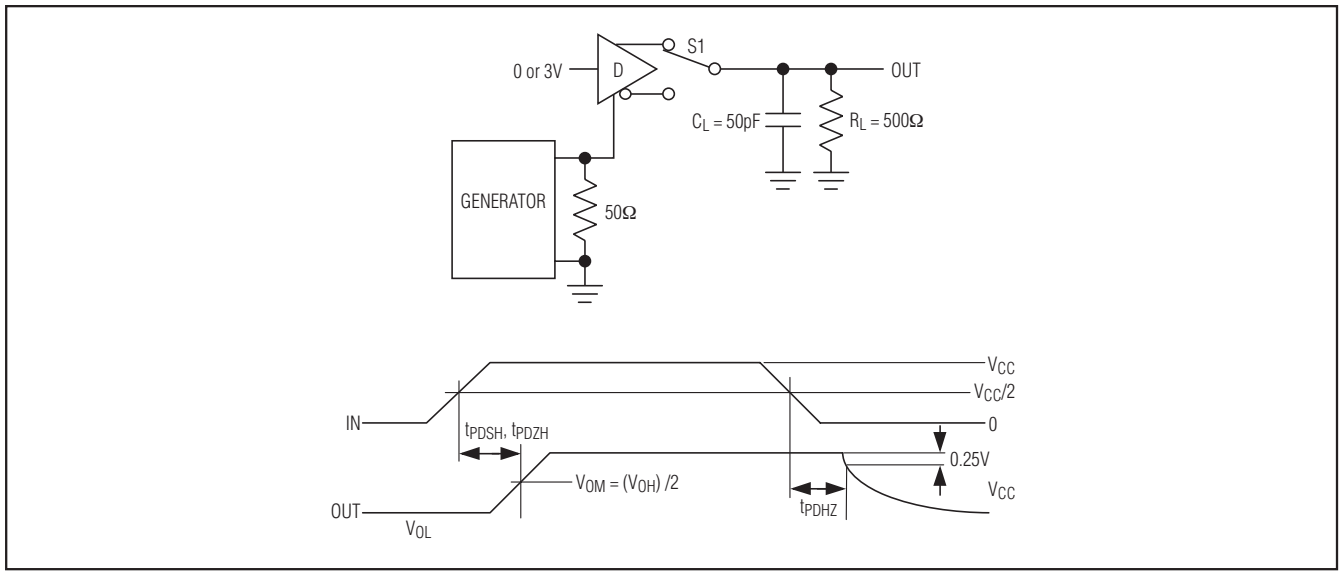


Figure 5. Driver Enable and Disable Times (t_{PDSH}, t_{PDZH}, t_{PDHS}, t_{PDHZ})

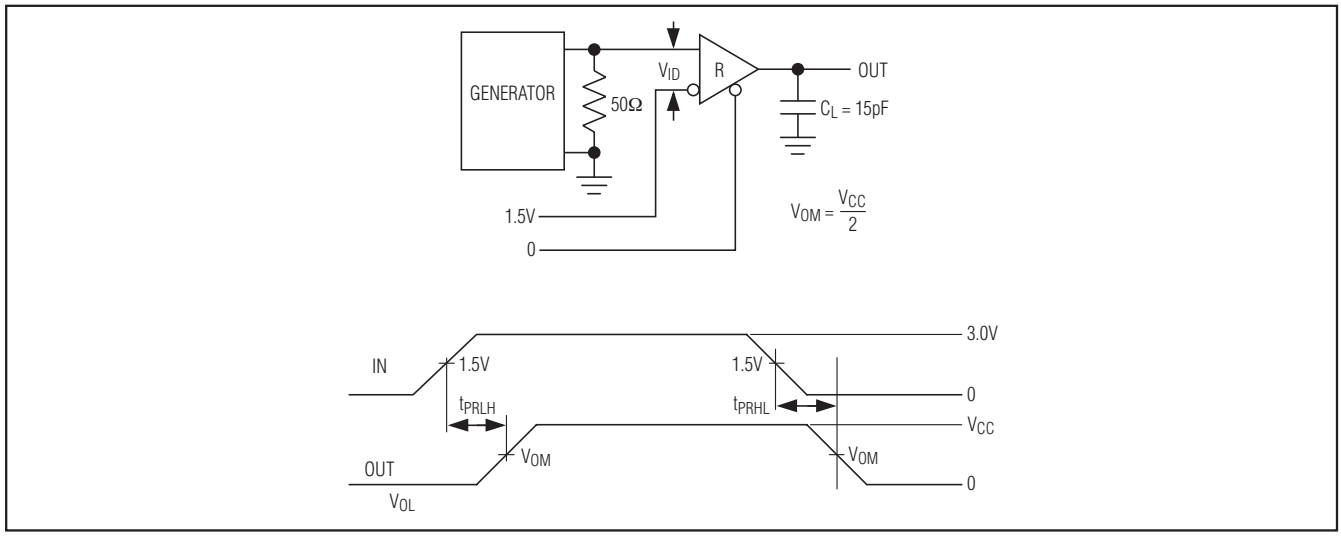


Figure 6. Receiver Propagation Delays

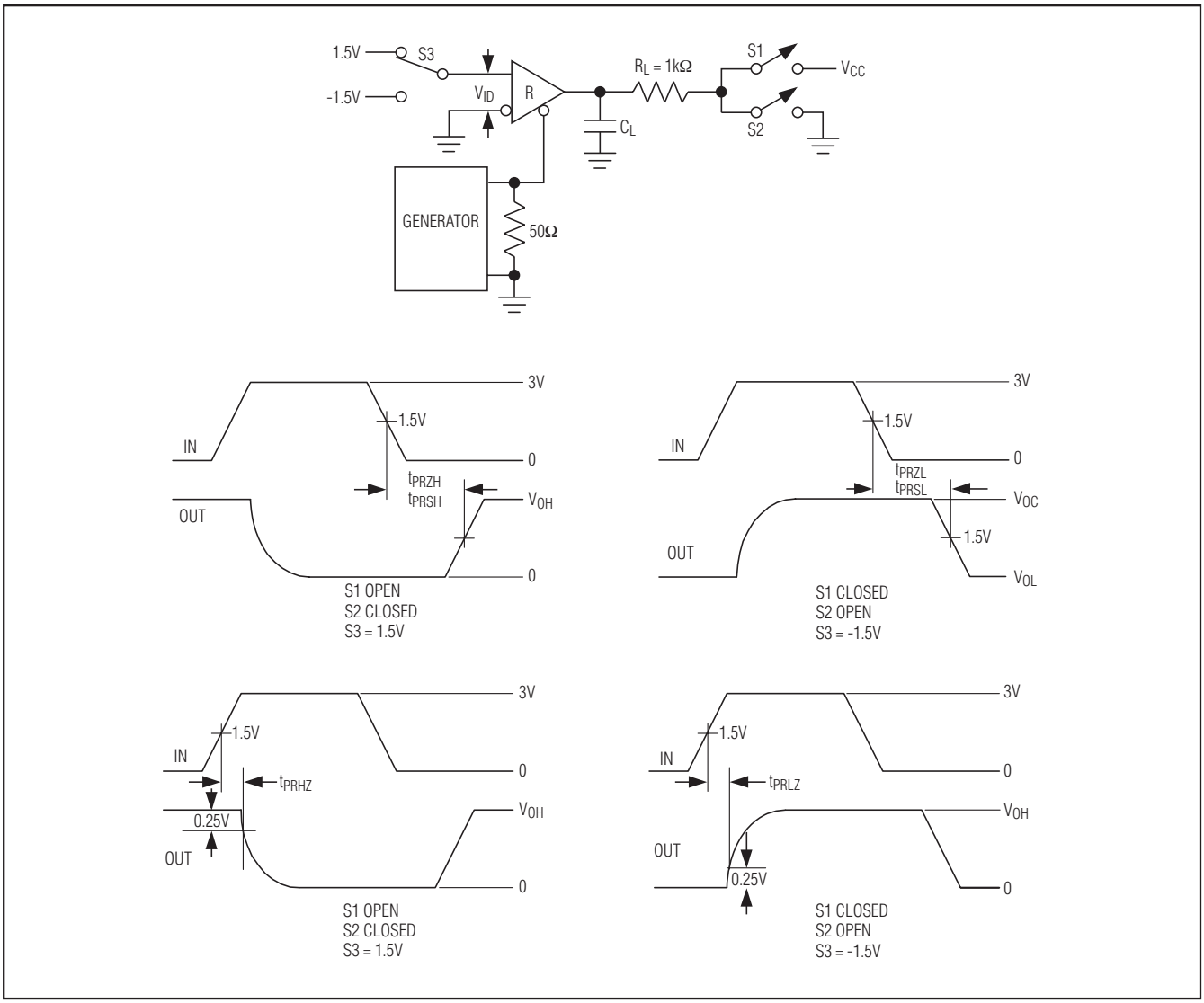


Figure 7. Receiver Enable and Disable Times

Detailed Description

The MAX3362 low-power, high-speed transceiver for RS-485/RS-422 communication operates from a single +3.3V power supply. The device contains one differential line driver and one differential line receiver. The driver and receiver may be independently enabled. When disabled, outputs enter a high-impedance state.

The transceiver guarantees data rates up to 20Mbps, with an output skew of less than 6ns. This low skew time makes the MAX3362 ideal for multidrop clock/data distribution applications, such as cellular base stations. Driver and receiver propagation delays are below 50ns. The output level is guaranteed at 1.5V on a standard 54Ω load.

The device has a hot-swap feature that eliminates false transitions on the data cable during circuit initialization. Also, drivers are short-circuit current limited and are protected against excessive power dissipation by thermal shutdown circuitry.

The MAX3362 has a 1/8-unit-load receiver input impedance, allowing up to 256 transceivers to be connected

Table 1. Transmitter Functional Table

TRANSMITTING				
INPUTS			OUTPUTS	
$\overline{RE}$	DE	DI	A	B
X	1	1	1	0
X	1	0	0	1
0	0	X	High Z	High Z
1	0	X	Shutdown	

Table 2. Receiver Functional Table

RECEIVING			
INPUTS			OUTPUT
$\overline{RE}$	DE	A – B	RO
0	X	≥ 200mV	1
0	X	≤ -200mV	0
1	1	X	High-Z
1	0	X	Shutdown

simultaneously on a bus. The MAX3362 is designed for half-duplex communication.

Driver

The driver transfers single-ended input (DI) to differential outputs (A, B). The driver enable (DE) input controls the driver. When DE is high, driver outputs are enabled. These outputs are high impedance when DE is low.

When the driver is enabled, setting DI low forces the noninverting output (A) low and inverting output (B) high. Conversely, drive DI high to force noninverting output high and inverting output low (Table 1).

Drive $\overline{RE}$ high and DE low (disable both receiver and driver outputs) to enter low-power shutdown mode.

Receiver

The receiver reads differential inputs from the bus lines (A, B) and transfers this data as a single-ended output (RO). The receiver enable ($\overline{RE}$) input controls the receiver. Drive $\overline{RE}$ low to enable the receiver. Driving $\overline{RE}$ high places RO into a high-impedance state.

When the receiver is enabled, RO is high if (A-B) ≥ 200mV. RO is low if (A-B) ≤ -200mV.

Drive $\overline{RE}$ high and DE low (disable both receiver and driver outputs) to enter low-power shutdown mode.

Hot-Swap Capability

Hot-Swap Input

When circuit boards are inserted into a hot or powered backplane, disturbances to the enable and differential receiver inputs can lead to data errors. Upon initial circuit board insertion, the processor undergoes its power-up sequence. During this period, the output drivers are high impedance and are unable to drive the DE input of the MAX3362 to a defined logic level. Leakage currents up to 10µA from the high-impedance output could cause DE to drift to an incorrect logic state. Additionally, parasitic circuit board capacitance could cause coupling of V_{CC} or GND to DE. These factors could improperly enable the driver.

When V_{CC} rises, an internal pulldown circuit holds DE low for at least 10µs and until the current into DE exceeds 200µA. After the initial power-up sequence, the pulldown circuit becomes transparent, resetting the hot-swap tolerable input.

Hot-Swap Input Circuitry

The MAX3362 enable inputs feature hot-swap capability. At the input there are two NMOS devices, M1 and M2 (Figure 8). When V_{CC} ramps from 0, an internal 10µs timer turns on M2 and sets the SR latch, which also turns on M1. Transistors M2, a 300µA current sink, and M1, a 30µA current sink, pull DE to GND through an 8kΩ resistor. M2 is designed to pull DE to the disabled state against an external parasitic capacitance up to 100pF that may drive DE high. After 10µs, the timer deactivates M2 while M1 remains on, holding DE low against three-state leakages that may drive DE high. M1 remains on until an external source overcomes the required input current. At this time, the SR latch resets and M1 turns off. When M1 turns off, DE reverts to a standard, high-impedance CMOS input. Whenever V_{CC} drops below 1V, the hot-swap input is reset.

For $\overline{RE}$ there is a complementary circuit employing two PMOS devices pulling $\overline{RE}$ to V_{CC}.

Hot-Swap Line Transient

The circuit of Figure 9 shows a typical offset termination used to guarantee a greater than 200mV offset when a line is not driven (the 50pF represents the minimum parasitic capacitance that would exist in a typical application). During a hot-swap event when the driver is connected to the line and is powered up the driver must not cause the differential signal to drop below 200mV. Figures 10, 11, and 12 show the results of the

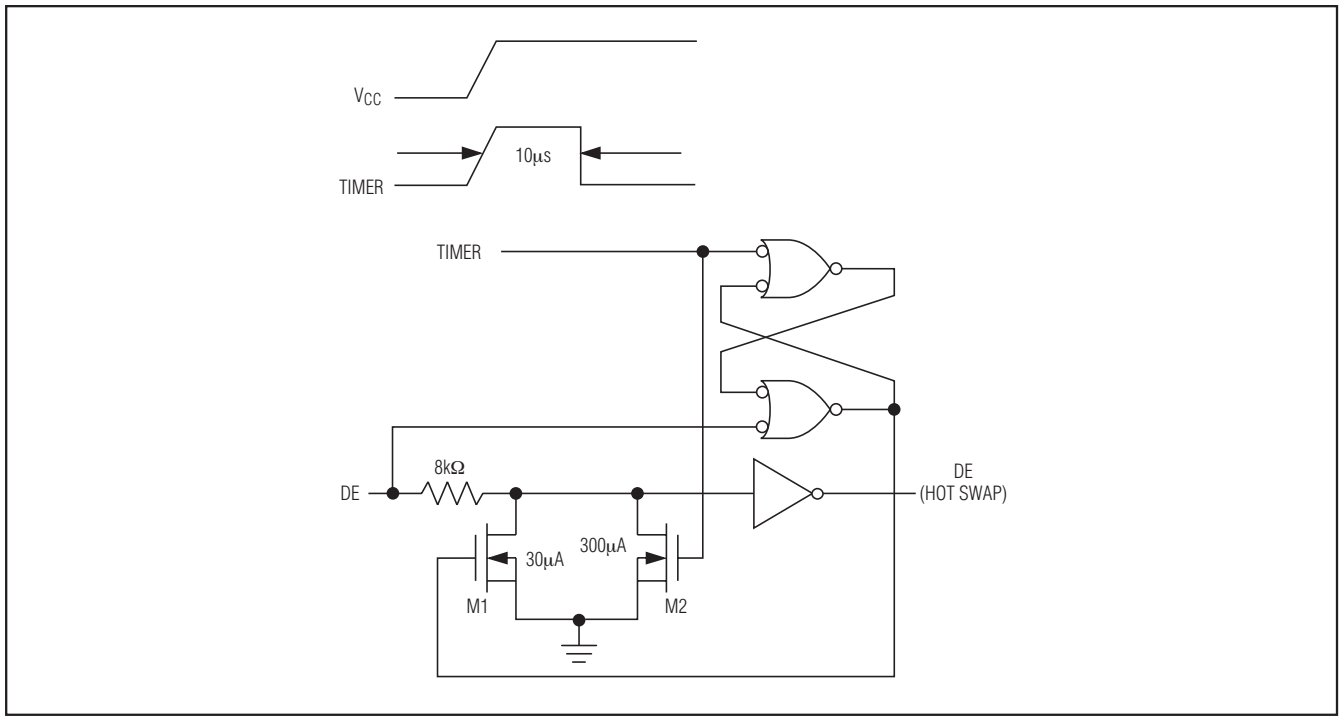


Figure 8. Simplified Structure of the Driver Enable Input (DE)

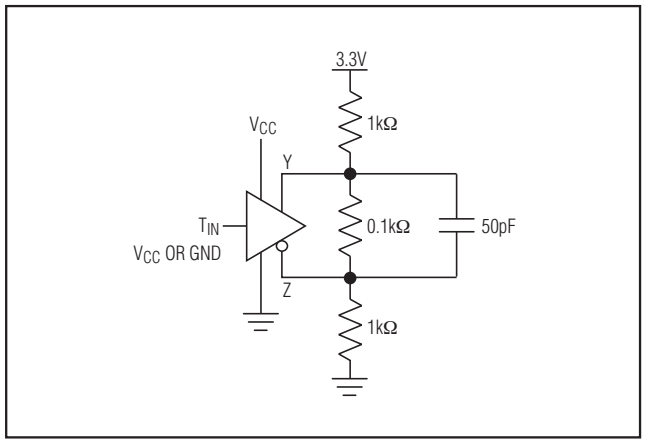


Figure 9. Differential Power-Up Glitch (Hot Swap)

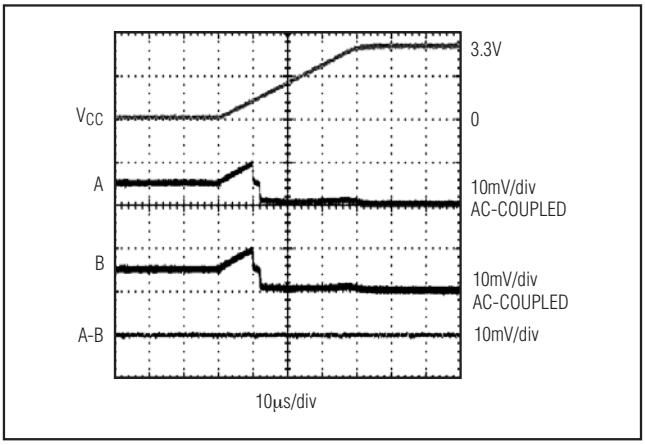


Figure 10. Differential Power-Up Glitch (0.1V/µs)

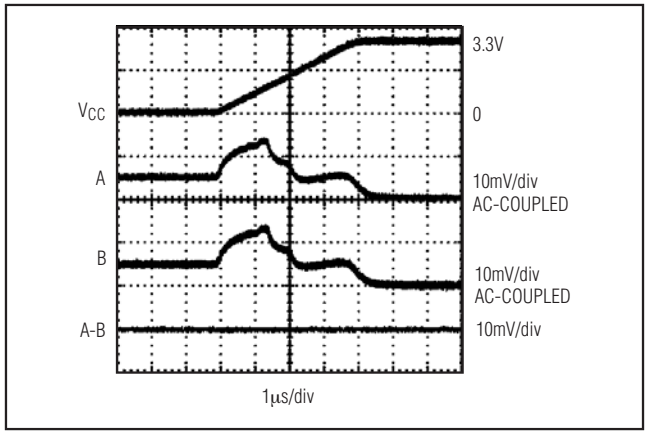


Figure 11. Differential Power-Up Glitch (1V/µs)

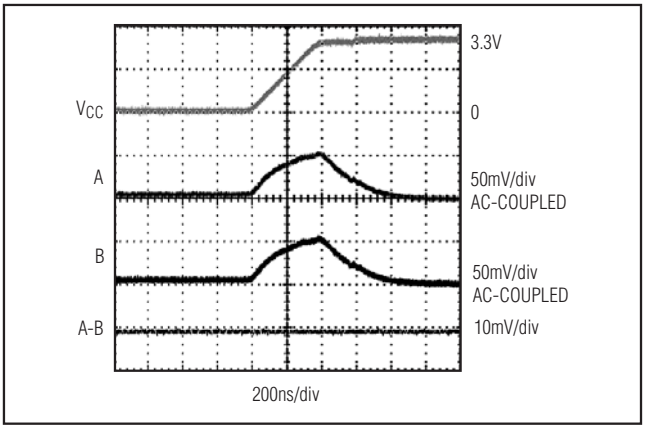


Figure 12. Differential Power-Up Glitch (10V/µs)

MAX3362 during power-up for three different V_{CC} ramp rates (0.1V/µs, 1V/µs, and 10V/µs). The photos show the V_{CC} ramp, the single-ended signal on each side of the 100Ω termination, as well as the differential signal across the termination.

Low-Power Shutdown Mode

Low-power shutdown mode is initiated by bringing both $\overline{RE}$ high and DE low. In shutdown, the MAX3362 typically draws only 1µA supply current.

$\overline{RE}$ and DE may be driven simultaneously; the device is guaranteed not to enter shutdown if $\overline{RE}$ is high and DE is low for less than 50ns. If the inputs are in this state for at least 600ns, the device will enter shutdown.

Enable times t_{PDZH}, t_{PDZL}, t_{PRZH} and t_{PRZL} in the *Switching Characteristics* table assume the device was not in a low-power shutdown state. Enable times t_{PDSH}, t_{PDSL}, t_{PRSH}, and t_{PRSL} assume the device was shut down. Drivers and receivers take longer to become enabled from low-power shutdown mode than from driver/receiver disable mode.

Applications Information

Propagation Delays

Figures 5 and 6 show the typical propagation delays. Skew time is simply the difference between the low-to-high and high-to-low propagation delay. Small driver/receiver skew times help maintain a symmetrical mark-space ratio (50% duty cycle). Both the receiver skew time and driver skew time are under 6ns.

256 Transceivers on the Bus

The standard RS-485 receiver input impedance is 12kΩ (one-unit load), and a standard driver can drive up to 32 unit loads. The MAX3362 transceiver has a 1/8-unit-load receiver input impedance (96kΩ), allowing up to 256 transceivers to be connected in parallel on one communication line. Any combination of these devices and/or other RS-485 transceivers with a total of 32 unit loads or less can be connected to the line.

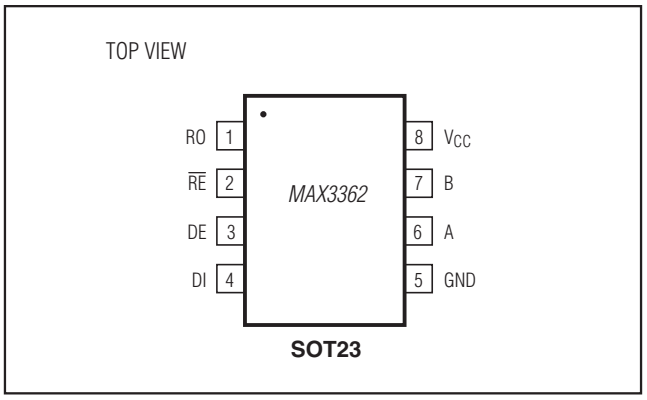
Driver Output Protection

Two mechanisms prevent excessive output current and power dissipation caused by faults or by bus contention. The first, a foldback current limit on the output stage, provides immediate protection against short circuits over the whole common-mode voltage range (see *Typical Operating Characteristics*). The second, a thermal shutdown circuit, forces the driver outputs into a high-impedance state if the die temperature becomes excessive.

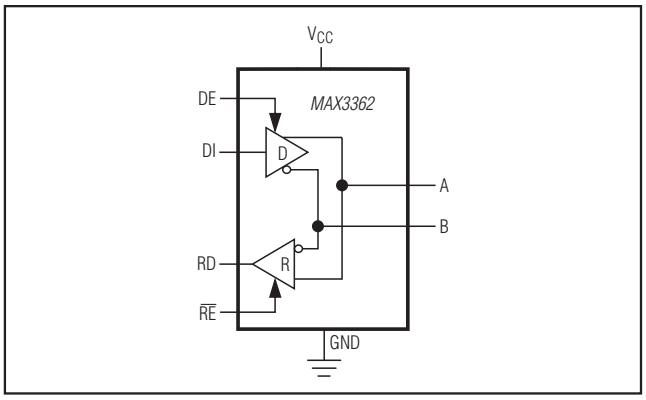
Typical Applications

The MAX3362 transceiver is designed for bidirectional data communications on multipoint bus transmission lines. The *Typical Operating Circuit* shows a typical network applications circuit. To minimize reflections, the line should be terminated at both ends in its characteristic impedance, and stub lengths off the main line should be kept as short as possible.

Pin Configuration



Functional Diagram

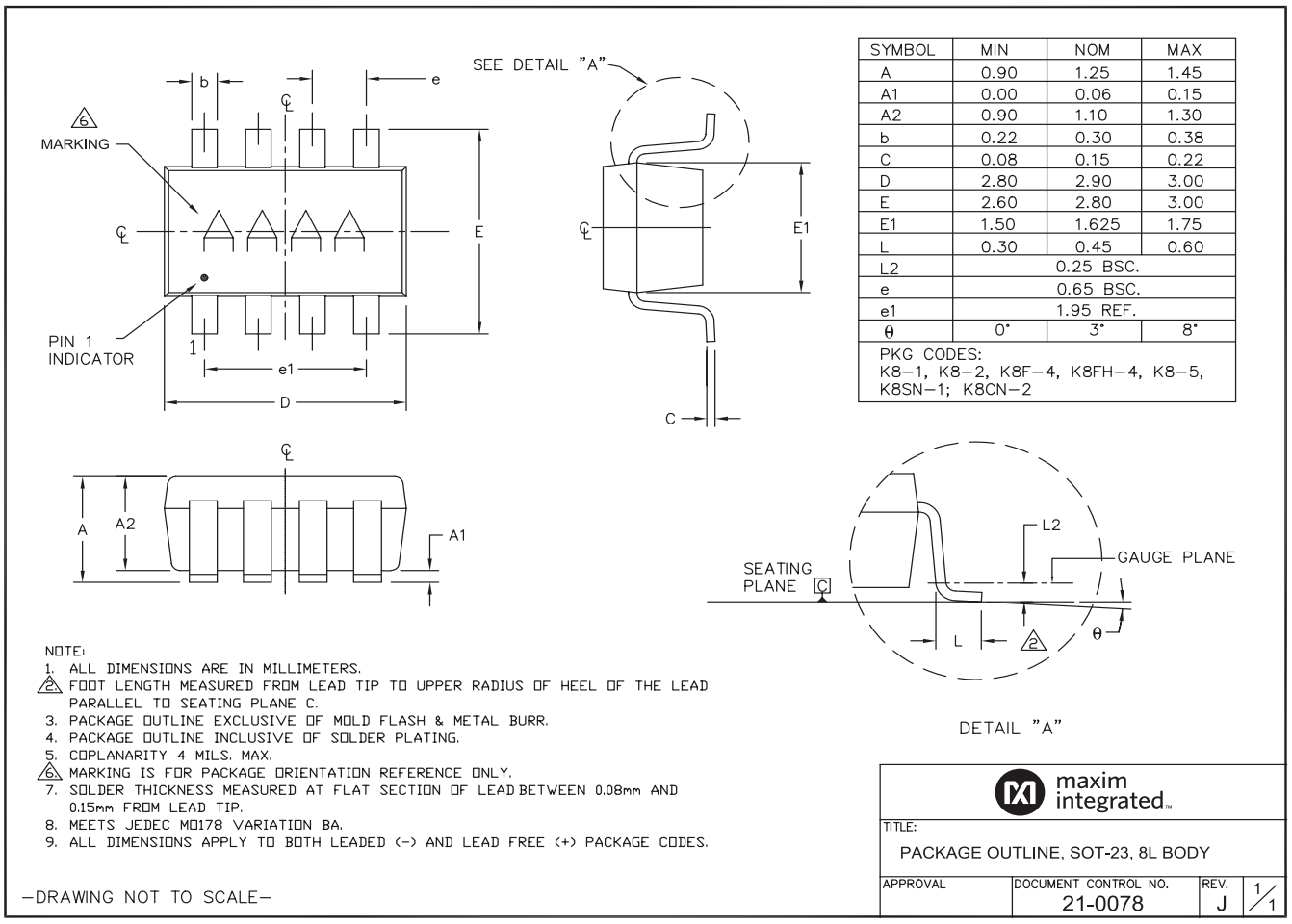


Chip Information

PROCESS: BiCMOS

Package Information

For the latest package outline information and land patterns (footprints), go to **www.maximintegrated.com/packages**. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.



SYMBOL	MIN	NOM	MAX
A	0.90	1.25	1.45
A1	0.00	0.06	0.15
A2	0.90	1.10	1.30
b	0.22	0.30	0.38
C	0.08	0.15	0.22
D	2.80	2.90	3.00
E	2.60	2.80	3.00
E1	1.50	1.625	1.75
L	0.30	0.45	0.60
L2	0.25 BSC.		
e	0.65 BSC.		
e1	1.95 REF.		
θ	0°	3°	8°

PKG CODES:
K8-1, K8-2, K8F-4, K8FH-4, K8-5, K8SN-1; K8CN-2

NOTE:
1. ALL DIMENSIONS ARE IN MILLIMETERS.
2. FOOT LENGTH MEASURED FROM LEAD TIP TO UPPER RADIUS OF HEEL OF THE LEAD PARALLEL TO SEATING PLANE C.
3. PACKAGE OUTLINE EXCLUSIVE OF MOLD FLASH & METAL BURR.
4. PACKAGE OUTLINE INCLUSIVE OF SOLDER PLATING.
5. COPLANARITY 4 MILS. MAX.
6. MARKING IS FOR PACKAGE ORIENTATION REFERENCE ONLY.
7. SOLDER THICKNESS MEASURED AT FLAT SECTION OF LEAD BETWEEN 0.08mm AND 0.15mm FROM LEAD TIP.
8. MEETS JEDEC MO178 VARIATION BA.
9. ALL DIMENSIONS APPLY TO BOTH LEADED (-) AND LEAD FREE (+) PACKAGE CODES.

-DRAWING NOT TO SCALE-

maxim integrated™

TITLE:
PACKAGE OUTLINE, SOT-23, 8L BODY

APPROVAL	DOCUMENT CONTROL NO.	REV.	
	21-0078	J	1/1

Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
4	2/15	Updated the *Benefits and Features* section	1

